

ECHO CANCELLING/SUPPRESSION FOR HANDSETS

Field Of The Invention

The present invention relates generally to communications and in particular to an echo suppressor and method for suppressing echoes in a communication path.

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Background Of The Invention

Acoustic coupling in telephone device handsets is common and occurs when signals broadcast by the speaker of a telephone device handset are coupled to the microphone of the same telephone device handset. This acoustic coupling results in echo signals being applied to the network over which the communication channel is established. The acoustic suppression of signals broadcast by the handset speaker before they are picked up by the handset microphone is variable and depends on a number of factors, such as the pressure exerted on the user's ear by the handset speaker and how well the handset speaker seals the ear. The loss in power of these signals as a result of acoustic suppression can vary from about -50db to as low as -24db when the telephone device handset is placed on a hard surface. If the delay in the network is short, such as for example 30ms, the normal acoustic suppression of signals broadcast by the handset speaker is generally sufficient to inhibit voice quality from being noticeably degraded by the echo signals picked up by the handset microphone. However, if the delay in the network is significant, echo signals applied to the network due to acoustic coupling will noticeably degrade voice quality over the communication channel. As such, suppressing echo signals resulting from acoustic coupling in telephone device handsets is important.

Echo cancelers in telephone device handsets have been considered. A typical echo canceler attempts to model the transfer function of the echo signal path using a linear algorithm such as a Least-Mean-Squared (LMS) algorithm. The estimated echo signals generated by the echo canceler are subtracted from the echo signals picked up by the handset microphone. Differences between the estimated echo signals and the actual echo signals result in error signals, which are fed back to the echo canceler. Unfortunately, since the algorithm executed by the echo canceler is linear, the echo canceler cannot deal with non-linear effects and can only converge to

a transfer function which approximates the echo signals. As a result, residual echo error signals are applied to the network.

It is therefore an object of the present invention to provide a novel echo suppressor and method for suppressing echoes in a communication path.

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Summary Of The Invention

According to one aspect of the present invention there is provided a method for suppressing echo signals generated in a communication path comprising the steps of:

10 monitoring signals supplied to said communication path to determine an attribute thereof; and

masking signals received from said communication path as a function of the determined attribute of said monitored signals.

In a preferred embodiment, the attribute is the power level of the
15 monitored signals. During the monitoring step, power level calculations are performed to determine the power level of the monitored signals. This is achieved by generating an envelope following the power level of the monitored signals. The envelope is generated by an infinite impulse response (IIR) lowpass filter. The IIR lowpass filter generates the envelope by solving the equation:

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$$\text{AbsY} = (1 - \alpha) \text{AbsY} + \alpha * \text{AbsY}_0$$

where alpha is a parameter of the IIR filter.

An echo signal value is then calculated by solving the equation:

$$\text{Echo} = \text{AbsY} / 10^{(A/20)}$$

where A is the minimum attenuation of echo signals in the communication path. The
25 calculated echo signal value is used to select a mask to be combined with digitized signals received from the communication path.

Preferably, the mask is a string of n-bits where n is a function of the echo signal value and wherein at least the most significant bits of the string have a zero value. In one embodiment, the bits of the mask are all zeros to basically achieve
30 total echo suppression. In an alternative embodiment, the mask is leaky and at least the least significant bit of the mask has a one value.

a power level calculator determining the power level of signals supplied to said communication path; and

According to still yet another aspect of the present invention there is provided in a telephone device including a handset having a speaker to broadcast incoming signals and a microphone to receive outgoing signals, an echo suppressor to suppress echo signals picked up by the microphone as a result of acoustic coupling between said speaker and microphone comprising:

a mask generator responsive to said power level calculator and generating masks, said masks being generated as a function of the determined power and being applied to said outgoing signals thereby to suppress echo signals.

an echo canceler in parallel with said communication path, said echo canceler having a transfer function approximating that of said communication path and generating estimated echo signals in response to signals supplied to said communication path, said echo canceler subtracting said estimated echo signals from signals received from said communication path to generate residual echo error signals; and

a processor receiving said estimated echo signals and said residual echo error signals, said processor including a power level calculator to determine the power level of the estimated echo signals; and a mask generator responsive to the power level calculator and generating masks, said masks being generated as a

function of the determined power level and being applied to said residual echo error signals thereby to suppress the same.

The present invention provides advantages in that since echo signals are adaptively masked by the echo suppressor, echo suppression is achieved while
5 maintaining voice quality.

Brief Description Of The Drawings

Embodiments of the present invention will now be described more fully with reference to the accompanying drawings in which:

10 Figure 1 is a schematic diagram of a communication channel established between a pair of telephone devices over a network showing acoustic coupling in one of the telephone devices;

Figure 2 is a schematic diagram of a telephone device including an echo suppressor in accordance with the present invention;

15 Figure 3 is a graph showing an envelope calculated by the echo suppressor of Figure 2 following the power level of a signal to be broadcast by a telephone device handset speaker;

Figure 4 is a graph showing an echo signal and the mask generated by the echo suppressor of Figure 2 to "cover" the echo signal picked up by a telephone
20 device microphone; and

Figure 5 is a schematic diagram of a conventional echo suppressor and a non-linear processor to suppress residual echo signals in accordance with the present invention.

Detailed Description Of The Preferred Embodiment

Turning now to Figure 1, a communication channel established between a pair of telephone devices 12 and 14 over a network 16 is shown and is generally indicated to by reference numeral 10. As can be seen, when a communication channel is established between the telephone devices 12 and 14,
30 acoustic signals 18 broadcast by the handset speaker 20 of receiving telephone device 14 are acoustically coupled to the handset microphone 22 of the telephone device 14.

The echo signals picked up by the handset microphone 22 as a result of the acoustic coupling cause echoes in the network 16, which degrade voice quality.

copy If the delay in the network 16 is long, such as for example 150ms, which may be caused by voice packetization and local area network (LAN) propagation delays, echoes in the network 16 as a result of acoustic coupling become audile detracting from voice quality.

To suppress adaptively echo signals picked up by the handset microphone 22 as a result of acoustic coupling, an echo suppressor 32 in accordance with the present invention is provided in each of the telephone devices 12 and 14 respectively. Figure 2 better illustrates telephone device 14. As can be seen, telephone device 14 includes a line receiver 40 coupled between the network 16 and an analog to digital converter 42. Analog to digital converter 42 provides digital output to the echo suppressor 32 and to a digital to analog converter 44. The digital to analog converter 44 conveys its analog output to a speaker driver 46, which amplifies the analog signals before they are broadcast by the handset speaker 20.

Coupled to the handset microphone 22 is a microphone receiver 48, which provides output to an analog to digital converter 50. Analog to digital converter 50 provides digital output to the echo suppressor 32. Echo suppressor 32 in turn supplies output to a digital to analog converter 52 that is coupled to the network 16 via a line transmitter 54.

As can be seen, echo suppressor 32 couples the handset speaker 20 and the handset microphone 22 of the telephone device 14. The echo suppressor 32 adaptively masks echo signals picked up by the handset microphone 22 to inhibit echo in the network 16. The adaptive masking performed by the echo suppressor 32 is based on the power level of signals to be broadcast by the handset speaker 20. This is due to the fact that typically, the larger the signals broadcast by the handset speaker 20, the larger the echo signals picked up by the handset microphone 22 as a result of acoustic coupling will be.

In the present embodiment, the echo suppressor 32 is embodied in a digital signal processor executing an echo suppression algorithm. The echo suppression algorithm performs a power level calculation 60 to determine the power level of signals received by the telephone device 14 to be broadcast by the handset

speaker 20 and uses the determined power level to generate masks. The masks are subtracted from signals received by the handset microphone 22 via a multiplier 62 to mask echo signals picked up by the handset microphone 22.

Appendix A shows psuedo-code representing the echo suppression algorithm executed by the echo suppressor 32. The echo suppression algorithm, in response to signals to be broadcast by the handset speaker 20, invokes a power level calculation routine (see Appendix B). During execution of this routine, an envelope following the power level of signals to be broadcast by the handset speaker is generated using an infinite impulse response (IIR) lowpass filter. The IIR filter generates the envelope by estimating the long-range average of the absolute value of the signal to be broadcast and is of the form:

$$\text{AbsY} = (1-\alpha)\text{AbsY} + \alpha * \text{AbsY0} \quad (1)$$

Alpha is an IIR filter parameter and is chosen to provide a fast attack time and a slow decay time for the IIR filter. In the present embodiment, two different values for alpha are used, namely alpha_fast and alpha_slow depending on the power level of the signal to be broadcast by the handset speaker 20. Figure 3 shows an example of an envelope 64 generated by the echo suppressor 32 in response to a signal to be broadcast by the handset speaker where alpha_fast = 1 and alpha_slow = 2⁻¹². As will be appreciated, by choosing these values for alpha, the echo suppressor generates an envelope that reacts fast to signals to be broadcast by the handset speaker 20. The slow decay time on the other hand compensates for small signal delays and reduces the switching effect when the signals fade.

As the envelope is generated, the echo suppressor 32 invokes a mask selection routine to calculate the maximum expected value of the echo signal based on the envelope. As stated earlier, the echo signal received by the handset microphone 22 is an attenuated copy of the signal broadcast by the handset speaker 20. The maximum expected value of the echo signal is calculated by solving the equation:

$$\text{Echo} = \text{AbsY} / 10^{(A/20)} \quad (2)$$

where A is the minimum attenuation or acoustic suppression of echo signals.

Thus, for example in a case where signals broadcast by the handset speaker 20 undergo a minimum attenuation of -24dB before being picked up as echo

signals by the handset microphone 22, the maximum expected value of the echo signal according to equation (2) is equal to $AbsY/15.8489$.

After the maximum expected value of the echo signal has been calculated, the mask selection routine selects the mask to be combined with the echo signal in accordance with the routine illustrated in Appendix B. In the present embodiment, the mask combined with the echo signal takes the form of a string of zeros n-bits long, where n is a function of the echo signal value determined at equation (2). The maximum value for n is determined by the maximum output value of the analog to digital converter 50, which in the present example is 8192. Solving equation (2) using this value for AbsY yields 517 which in binary format is 1000000100. As a result, a mask having ten zeros (i.e. $n = 10$) is required to mask this binary value.

For example, if the power level of the signal to be broadcast by the handset speaker 20 is equal to 1379 after solving equation (1) to determine AbsY, by solving equation (2) the expected echo signal level equals 87 (assuming $A = -24$ dB) which in binary format is 1010111. In accordance with the mask selection routine, a mask having seven zeros is chosen and is combined with the signals received by the handset microphone 22 and digitized by the analog to digital converter 50.

Figure 4 shows an echo signal 66 in a telephone device having a minimum acoustic attenuation or suppression equal to -24 dB together with the mask 68 selected by the echo suppressor 32. In the first half of the graph, it can be seen that the mask 68 completely covers the echo signal. In the second half of the graph, voice signals picked up by the handset microphone 22 superimposed on the echo signal are shown. As will be appreciated, the mask is orders of magnitude smaller than the voice signals. As a result, the mask causes only a minimum loss of speech quality.

As will be appreciated, the echo suppressor 32, by selecting masks having the appropriate number of zeros, adaptively masks echo signals to achieve basically total echo suppression.

If desired, "leaky" masks can be used to mask echo signals to inhibit noticeable switching, which may occur during total echo suppression in the presence of high background noise. "Leaky" refers to a mask having at least one least significant bit (LSB) with a "one" value. Generally, the number of LSBs having

"one" values is chosen depending on the number of bits in the selected mask. For example, if an 8-bit mask is generated the three LSBs of the mask can be leaked (i.e. have "one" values). If a 7-bit mask is generated, the two LSBs of the mask can be leaked. As will be appreciated, by leaking some of the background noise, switching is
5 reduced.

Turning now to Figure 5, an alternative embodiment of the present invention is shown. In this embodiment, a conventional echo canceler 70 is in parallel with a communication path 71 in which echo signals are generated. The communication path 71 may be a telephone device handset as described in the
10 previous embodiment. Alternatively, the communication path 71 may be a handsfree telephone, network which causes network reflections or other source of echo signals. The echo canceler 70 attempts to model the transfer function of the echo signal path using a Least-Mean-Squared (LMS) algorithm so that the echo canceler generates estimated echo signals that are the same as the echo signals received from
15 communication path 71. The estimated echo signals are subtracted 72 from the actual echo signals in an attempt to cancel the echo signals. The amount of cancellation is commonly referred to as ERLE. Differences between the estimated echo signals and the actual echo signals result in error signals. The error signals are fed back to the echo canceler 70 so that the echo canceler can attempt to converge to the correct
20 transfer function.

Unfortunately, the LMS algorithm only models linear effects in the echo path and does not deal with non-linear effects caused by for example, clipping, telephone key rattling, and frequency shifts. Therefore, the echo canceler converges to a transfer function that approximates the correct transfer function resulting in non-
25 zero error signals. This causes residual echo signals.

To suppress the residual echo signals, a non-linear processor (NLP) 80 is provided and receives the estimated echo signals output by the echo canceler 70 as well as the residual error signals output by the subtractor 72. The NLP 80 executes an echo suppression algorithm similar to that executed by echo suppressor 32 to
30 determine the power level of the estimated echo signals output by echo canceler 70 and to generate masks based on the determined power level. The masks are combined with the error signals to suppress the residual echo signals.

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